

Hardware Implementation of MLP and RBF Neural Networks onto Multiple Processing Nodes

Tejas Dalal

*Dhirubhai Ambani Institute
of Information and Communication Technology,
Gandhinagar, India*

Mazad Zaveri

*Dhirubhai Ambani Institute
of Information and Communication Technology,
Gandhinagar, India*

Abstract — In this paper, describe an architecture consisting of multiple Processing Nodes (PNs) for emulating/implementing Artificial Neural Networks (ANNs). Our architecture can be configured for implementing two types of ANNs (MLP and RBF). Our architecture has a connection memory that can be configured, in terms of the number of layers, and in terms of the inter-layer connectivity requirements. When compared to the existing CNAPS architecture, our architecture does not require the inter-PN buses, leading to a reduction in the communication time, and the area. Our architecture is based on $\langle 8, 6 \rangle$ fixed-point number representation. Our architecture (applied to an example recognition task using Iris dataset) has been implemented/synthesized on two 45 nm technology based platforms (ASIC and FPGA), and results have been provided, indicating that an ASIC implementation is more advantageous in terms of area/speed/power.

I. INTRODUCTION

The human cortex is a complex and fast computing machine. Memory, thought process, problem solving, and decision making capabilities of the cortex have inspired researchers to invent mathematical/computing models that are able to do similar operations, and are referred to as Artificial Neural Networks (ANNs). Such models capture the cortex's function at the most fundamental level (neural-level of abstraction [1]). A neuron is a basic cellular unit of the cortex, and each neuron can be approximated as a simple processing unit, which receives and combines several signals. The basic model of an artificial neuron is shown in fig.1.

The generalized architecture of ANN, generally consists of several Processing Nodes (PNs), where each PN could emulate one or more neurons [1]. A PN (emulating one neuron) has several inputs, and combines it usually by a simple summation. The output of a PN can be the input of other PNs through weighted connections, which correspond to the strength of synapses.

The most challenging applications of ANNs are: hand written character recognition, function approximation, time series prediction, control systems, etc.

II. ARCHITECTURE OF ANN

The architecture of ANN is shown in fig.2, where, several PNs are connected via shared bus through a control block, which takes all decisions, required for communication between PNs. Our proposed system architecture is based on the regional-broadcast scheme suggested in [1], and the ANN implementations in [2]. The internal architecture of each PN is shown in fig.3.

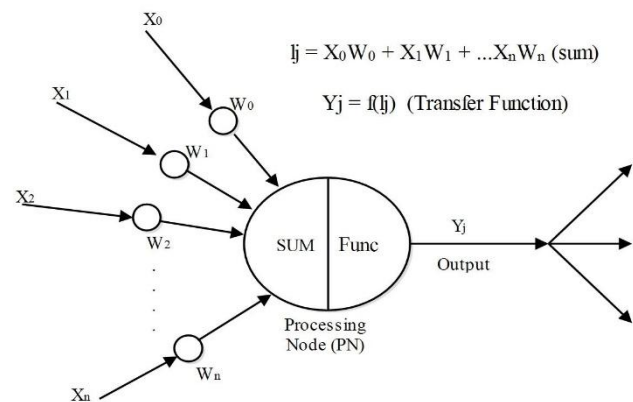


Fig. 1. Architecture of Artificial Neural Network

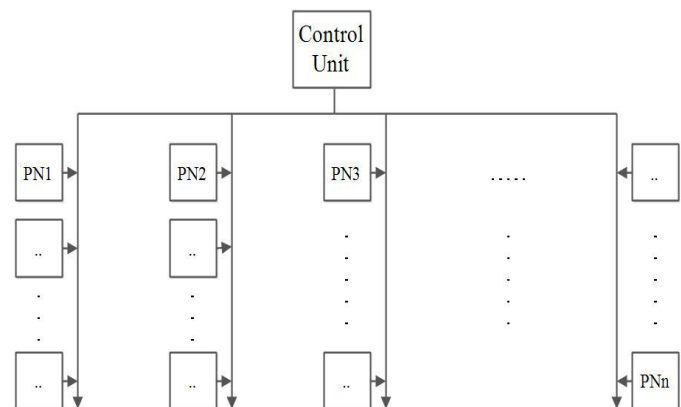


Fig. 2. Architecture of ANN

A. Activation Functions

Two activation functions have been implemented:

- 1) Sigmoid function for Multilayer perception (MLP) neural network
- 2) Gaussian function for Radial Basis Function (RBF) neural network

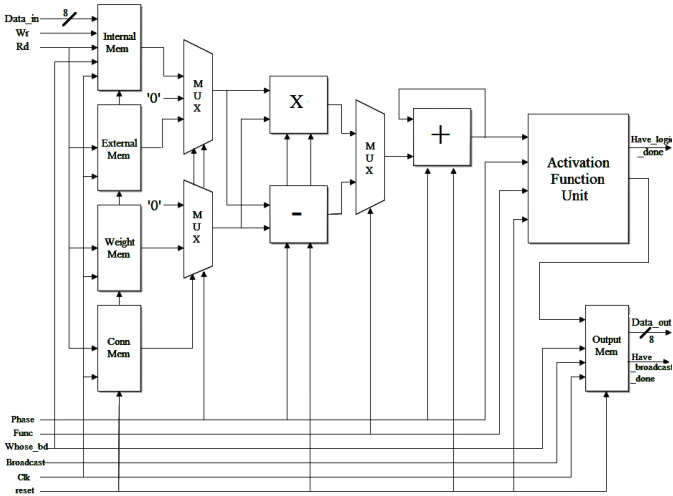


Fig. 3. Architecture of PN

The former function is a logistic function that ranges from 0 to 1, and is similar to hyperbolic tangent function, which ranges from -1 to 1. Therefore the input range is bounded in the range of [-2, 2] and that can be represented using <8, 6> fixed-point representation, where 2 bits are needed for the integer part and remaining 6 bits are for fraction part, providing an accuracy of approximately 0.0015625.

1) *Sigmoid (activation) function for Multilayer perception (MLP) neural network*: The architecture of activation function for MLP network is shown in fig.4. The Exponential function can be implemented, in hardware, using Piecewise Linear approximation (PWL) algorithm [3], [4]. The multiplication and the exponential function can be approximated using shifter and adder as shown in Table I. One output is chosen from the three different shifters which are controlled by the range of the input.

TABLE I
IMPLEMENTATION METHOD

operation	condition	Flags			
		z1	z2	z3	z4
$Y = 1$	$\ X\ \leq 5$	0	0	0	1
$Y = 0.03125 * \ X\ + 0.84375$	$2.375 < \ X\ < 5$	0	0	1	0
$Y = 0.125 * \ X\ + 0.625$	$1 < \ X\ < 2.375$	0	1	0	0
$Y = 0.25 * \ X\ + 0.5$	$0 < \ X\ < 1$	1	0	0	1
$Y = 1 - Y$	$\ X\ < 0$				

2) *Gaussian (activation) function for Radial Basis Function (RBF) neural network*: The Gaussian function for the RBF network has been implemented using Look up Table (LUT). The input to this function is the distance of the input

vector from the stored center. This distance can be linearly

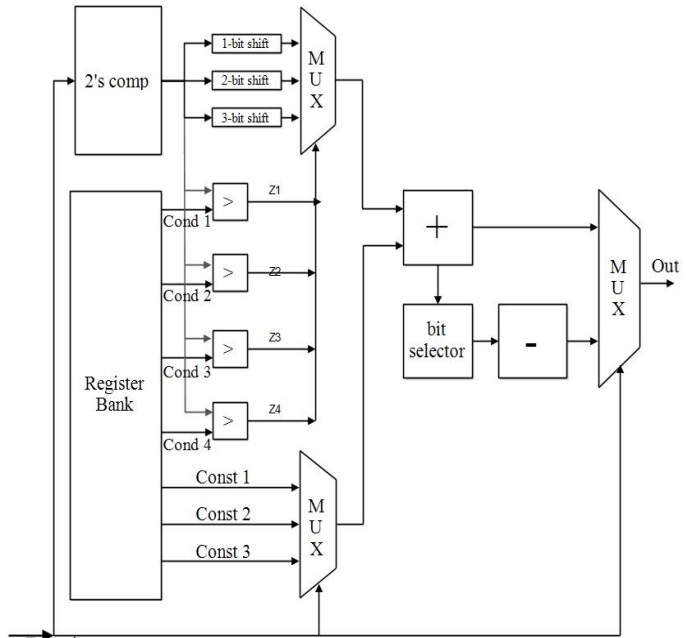


Fig. 4. Architecture of activation function for MLP network

mapped to the output.

The input to the activation function can be treated as the address for the LUT. The LUT contains the output corresponding to the input (or address), which is actually obtained by discretizing the Gaussian curve, into (output, input) pairs. As the range of the RBF output is bounded in the range of [0, 1], it is easy to make such table, by simply dividing by 2^{width} .

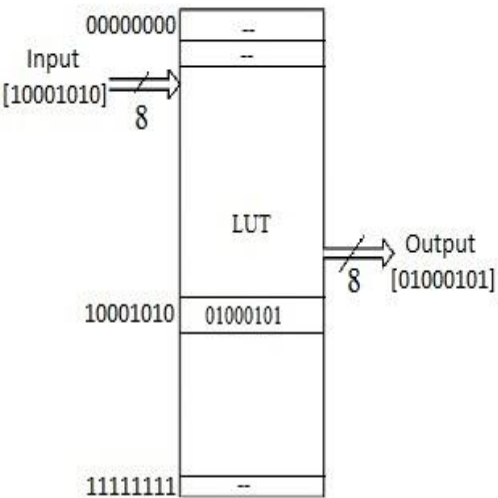


Fig. 5. Architecture of activation function for RBF network

For 8 bit width, the difference between two subsequent elements of the table will be $1/256 = 0.0039$; in other words, the range [0, 1] of the output is divided in 256 steps.

III. EXISTING ARCHITECTURES OF ANN

The approach of using (single) PN was introduced in ANNA chips [5]. The architecture of ANNA chips is shown in fig.6. Later on, multi-PN approach was shown in CNAPS [6], which consisted of 65 PNs. The architecture of CNAPS is shown in fig.7.

A. ANNA Chips

ANNA stands for Analog Neural Network Arithmetic and logic unit. A neural network for handwritten digit recognition was implemented with 136000 synapses on a mixed analog/digital chip. It was reported that, ANNA chips could recognize 1000 characters per second, with an error rate of 5% with floating point precision.

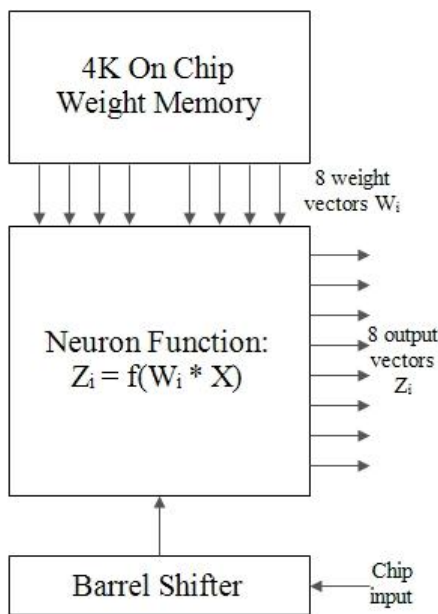


Fig. 6. Architecture of ANNA chips

ANNA chips was implemented on a 0.9 μm CMOS technology. It implemented 4096 physical synapses, which could be multiplexed. Resolution of the weights/synapses was 6 bit, and that of the input/output was 3 bit. All inputs/outputs of the chip were digital, while the internal computation technique was analog. The optical character recognition (OCR) algorithm required four layers, and each layer was computed sequentially.

B. CNAPS Architecture

CNAPS architecture had an array of processors (or PNs). These PNs were interconnected by three buses:-

- 1) *PNCMD bus*: broadcast the instruction from the sequencer to the PNs.
- 2) *IN/OUT bus*: broadcast data from one PN to all the PNs (8 bit wide).
- 3) *INTER-PN bus*: transfer data from each PN to one of its nearest neighbors (2 bit wide).

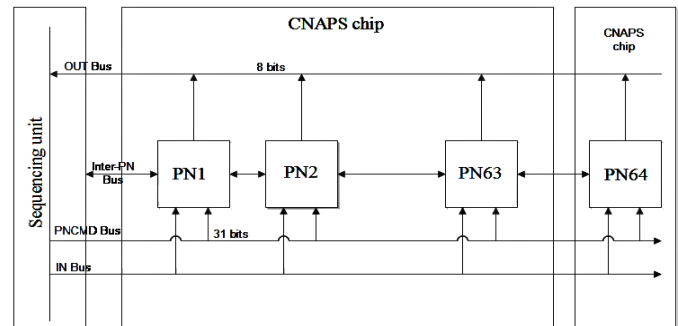


Fig. 7. Architecture of CNAPS

CNAPS network has layers, which are 1D or 2D structures. Two 2D layers could be connected locally. A 2D and a 1D layer, or two 1D layers could be fully connected. A simulation was reported on CNAPS with 128 PNs, working at 20 MHz clock.

C. Comparison of the Architectures

Comparison of our proposed architecture with the conventional architectures (ANNA chips, CNAPS) can be based on the following aspects:-

- 1) *Topology* - Both ANNA and CNAPS architectures have fixed topologies. In our architecture, one connection memory was introduced, which can be configured to implement different topologies.
- 2) *Requirement of buses* - In our architecture, all PNs communicate to each other only through the shared bus via control unit. Hence our architecture does not require the inter-PN bus (required in CNAPS).
- 3) *Type of networks implemented* - Both ANNA and CNAPS architectures have a single activation function and network (i.e. MLP neural network), whereas in our architecture the activation function and the network is configurable as either MLP or RPF neural network.
- 4) *Type of chip that used* - ANNA chips was based on mixed analog/digital signals, whereas CNAPS and our architecture are based on fully digital signals.

IV. FUNCTIONAL VERIFICATION BY IMPLEMENTING AN EXAMPLE

A recognition/classification example based on Iris dataset was implemented on our architecture using 4 PNs, as shown in fig.8. This example, (as shown in fig. 9) requires 12 neurons in the hidden layer and 3 neurons in the output layer. A total of four PNs were assumed, hence each PN would have 3 neurons in hidden layer (indicated by phase 0 in connection memory) and 1 neuron in output layer (indicated by phase 1 in connection memory).

All 15 (=12 + 3) neurons have been equally distributed (as shown in fig. 10), across the PNs. Each PN has the same number of neurons to be emulated, and hence, takes the same amount of time for computation (for each layer), eliminating the need for any PN to wait, before the communication starts.

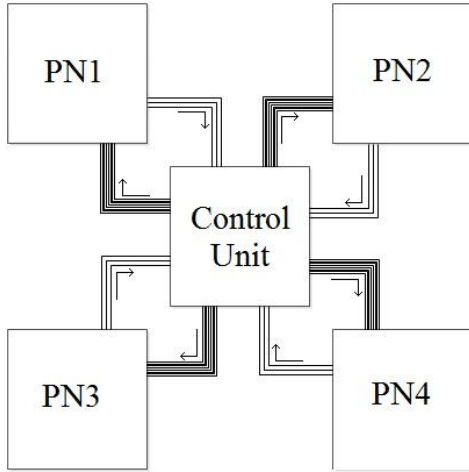


Fig. 8. Implementation of the Iris example, using 4 PNs

As shown in fig.9, the number of inputs required is 4 as per Iris dataset. As the network is fully connected, all neurons will get the same inputs, but this connection scheme, if required, can be configured by setting the appropriate bits in the connection memory.

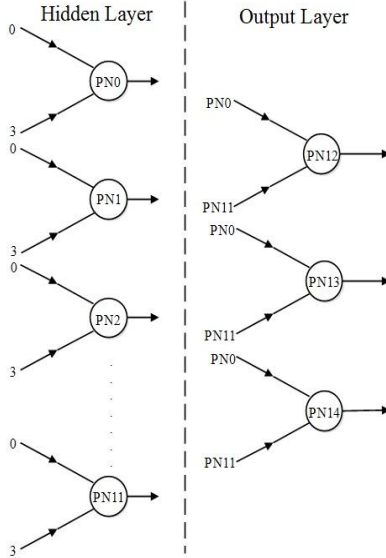


Fig. 9. Example

A. Calculations

Area required (for ASIC implementation) for each PN is 71439 m^2 (obtained from Cadence tool, refer to Table III) = $270 \text{ m} \times 270 \text{ m}$ {length x width}. So if one (additional) PN is attached to the architecture (for scaling-up the architecture), then the additional length requirement of the communication bus (= 1) is approximately 270 m.

The delay changes as length of wire (bus) increases, i.e. number of PN increases. The wire delay has nonlinear relation to the length [7] that is:

$$\tau_{wire} = (2.3 * R_0 + Rl)cL = \left(\frac{2.3 * R_0}{Rl} + 1 \right) Rcl^2 \quad (1)$$

Where,

R_0 = Channel resistance = $4 \text{ k}\Omega$

R = Wire resistance = 0.08Ω {for metal layers M1 to M3}

c = Wire capacitance = $4.5\text{--}3.5 \text{ aF}/\lambda$ {for metal layers M1 to M3}

L = Wire length = $270 \mu\text{m}$

So, total communication time,

$$T_{communication} = ((\tau_{wire} + \tau_{inv}) * N_{PNs}) * N_{neuron} \quad (2)$$

Where,

τ_{inv} = Inverter delay = $f + p = gh + p$

N_{PNs} = Total number of PNs = 1 to 128 (varying)

N_{neuron} = Total number of neurons = 10,000 (assumed)

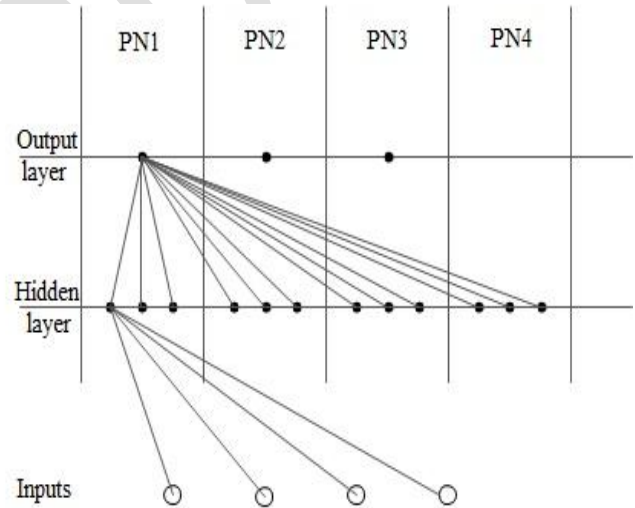


Fig. 10. Neuron distribution scheme for the Iris example

As the wire delay is in the range of picoseconds and inverter(repeater) delay is in range of few nanoseconds, the overall effect of the repeater and increased length for the wire doesn't affect the total communication time because the time for communication is in the range of hundreds of nanoseconds as seen in fig.11. In other words, the communication time is proportional to the number of neurons per PN, and only negligibly proportional to the total number of PNs (see [8] for more details).

The communication delay is almost linearly increasing, with increasing number of PNs; the computation delay is exponentially decreasing with increasing number of PNs (refer to fig.12). So, total time (required for completing all operation of ANN) decreases as the number of PNs increases and saturates (becomes dependent on communication delay) after a certain number of PNs as seen in fig.13.

B. Results and Discussions

70% of the Iris dataset was used for ANN training, and testing was done on the remaining 30%. For the Iris example, the estimated total simulation time (computation + communication delay) = $12 + 12 + 12 = 36$ cycles, where cycle time is 50ns (corresponding to frequency of 20MHz). So, simulation time = $36 \times 50\text{ns} = 1800\text{ns}$, and Total # of Connections (synapses) $12 \times 4 + 3 \times 12 = 84$. So the speed, in terms of Connections Per Second (CPS), of our architecture is $64 / 1800 \text{ ns} = 46.66 \text{ MCPS}$.

1) **Matlab® Results:** NNtool in Matlab® was used to obtain the trained weights for both the (MLP and RBF) neural networks. Training was done for 1000 iterations, with an error in the range of 10^{-6} . The results that were obtained from the Matlab® were converted into <8, 6>-fixed-point representation, for comparing them to the results obtained from the behavioral and synthesized HDL implementations.

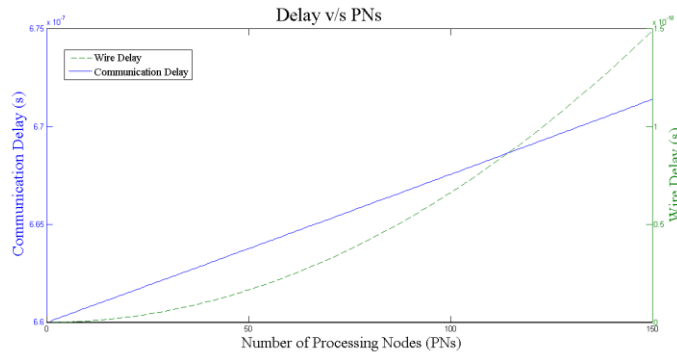


Fig. 11. Wire delay v/s number of PNs

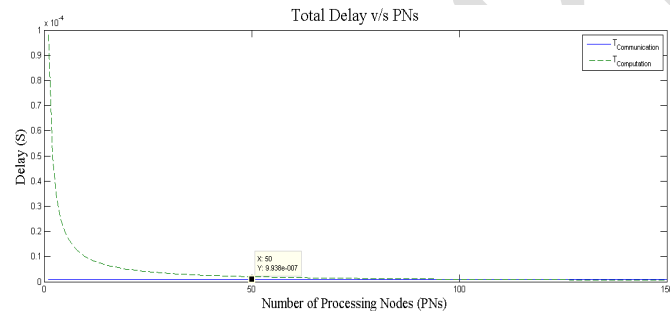


Fig. 12. Computation and communication delay v/s number of PNs

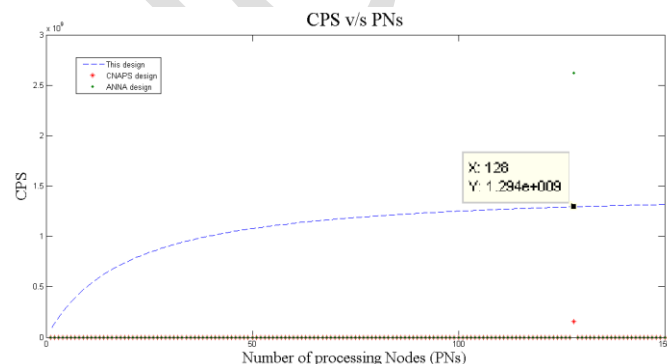


Fig. 13. Comparison of speed (ANNA chips, CNAPS, our architecture) different number of PNs

2) **FPGA Results:** The HDL code for the Iris example, was successfully synthesized using Xilinx tool, for a Spartan 6 FPGA board (45 nm technology). This implementation is capable of performing with the speed of 131.68 MCPS for an architecture with 4 PNs. If the number of PNs (running in parallel) increase, the speed can be further improved

3) **Cadence Results:** The result has been obtained for 45nm technology with nangate opencell slow library. For synthesis, RTL compiler tool was used, and for physical layout/designing encounter tool was used. The results, obtained from Cadence tools are given in Table II, Table III, and Table IV. Also final layout is shown in fig. 15.

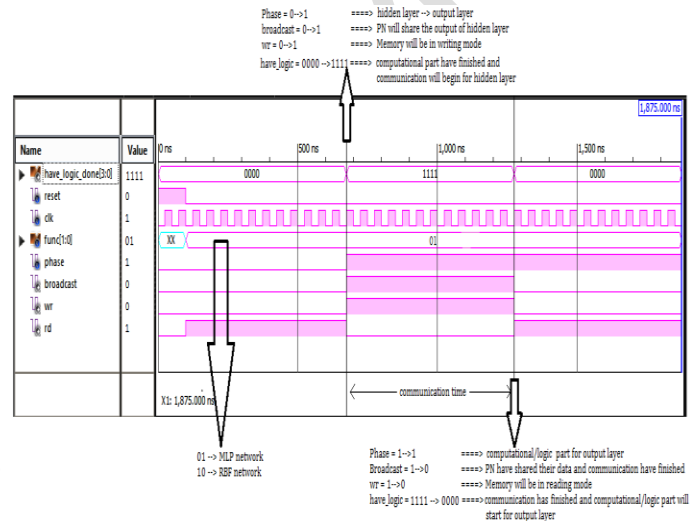


Fig. 14. FPGA functional verification

TABLE II
AREA RESULT FROM CADENCE

Type	Area (m ²)	Area (%)
Cell Area	101646	35.89 %
Net Area	181566	64.11 %
Total Area	283211	100 %

TABLE III
POWER RESULT FROM CADENCE

Type	Power (mW)	Power (%)
Leakage Power	1.286 mW	30.30 %
Dynamic Power	2.958 mW	69.70 %
Total Power	4.243 mW	100 %

TABLE IV
RESOURCE UTILIZATION RESULT FROM CADENCE

Type	Instances	Area (um ²)	Area (%)
Sequential	23272	68579.06	67.46
Inverters	5823	3220.20	3.20
Logic	22058	29856.26	29.35
Total	51153	101645.52	100
Average Fan-out	2.0		

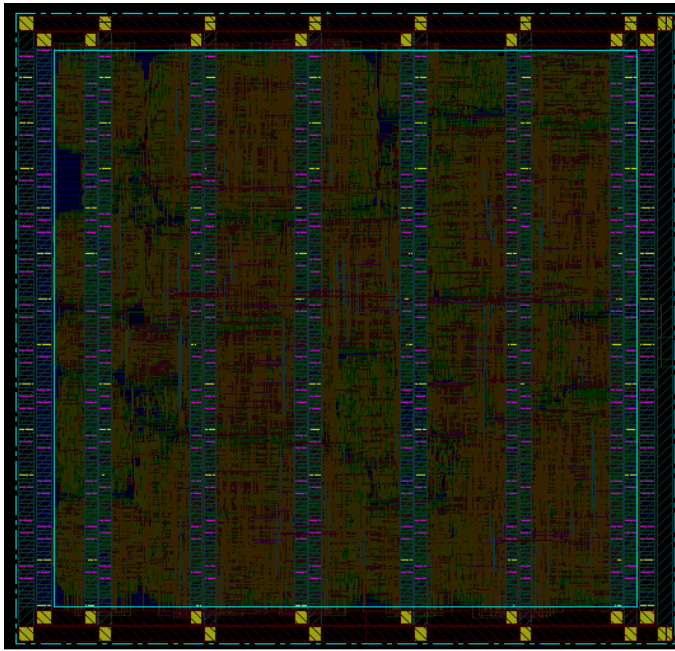


Fig. 15. ASIC Layout

4) *Comparison of Results with the Existing Architectures of ANN*: When compared to CNAPS, our architecture (which has a connection memory) provided more speed without extra cost in terms of area (i.e. inter-processor buses in CNAPS). Data displayed in the Table V has been adapted (and appropriately scaled) from [5] and [6]. From Table V, our architecture is 2x faster than CNAPS, while slower than ANNA chips (for obvious reason, that in ANNA chips, the multiply-summation operation is happening in parallel in the analog domain). The results given in Table VI, are for 4 PNs, so if more PNs run in parallel, then it is obvious that the speed advantage will be much more.

5) *Comparison of the ASIC and FPGA platform*: Comparison of the results for the two platforms are shown in Table VII, which shows that the ASIC design is much faster, less power consuming and requires less area, but the drawback of ASIC is high designing/manufacturing cost.

TABLE V
COMPARISON OF RESULTS

	Speed (MCPS)	Activation Function Implemented
CNAPS (128-PNs) 20MHz	153.2*	MLP
ANNA Chip	2624.0*	MLP
This Architecture (4-PNs) 20MHz	46.66	MLP, RBF
This Architecture (128-PNs) 20MHz	1294.33	

* values with appropriate technology scaling

TABLE VI
COMPARING TWO PLATFORMS

	ASIC (Cadence)	FPGA platform
Power	5.36 mW	26 mW
Speed	353.53 MCPS	171.56 MCPS
Area	283211 μm^2	1280000 μm^2

We can conclude that, the ASIC design can be implemented for very larger-scale architectures, where power, speed and area have higher priority, and when extremely large number of PNs are required (justifying the mass production and related manufacturing costs). FPGA can be used as an alternative to ASIC for small-scale architectures, where re-configurability has a higher priority than power, speed or area.

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